# EE 508 Lecture 16

# **Filter Transformations**

Lowpass to Highpass Lowpass to Band-reject **Review from Last Time**

#### **Standard LP to BP Transformation** 2 N  $s^2+1$ s s•BW  $\rightarrow$

- Standard LP to BP transform is a variable mapping transform
- $-$  Maps j $\omega$  axis to j $\omega$  axis
- Maps LP poles to BP poles
- Preserves basic shape but warps frequency axis
- Doubles order
- Pole Q of resultant band-pass functions can be very large for narrow pass-band
- Sequencing of frequency scaling and transformation does not affect final function

#### **Standard LP to BP Transformation Review from Last Time**



#### **Standard LP to BP Transformation Review from Last Time**

Frequency and s-domain Mappings - Denormalized

(subscript variable in LP approximation for notational convenience)



All three approaches give same approximation

Which is most practical to use? Often none of them !

#### **Standard LP to BP Transformation Review from Last Time**

Pole Mappings



Note doubling of poles, addition of zeros, and likely Q enhancement

## **LP to BP Transformation**

#### Pole Q of BP Approximations



#### **LP to BP Transformation**

Pole locations vs 
$$
Q_{LP}
$$
 and  $\delta$   

$$
\delta = \left(\frac{BW}{\omega_{M}}\right) \omega_{0LP}
$$



### **LP to BP Transformation**



#### Classical BP Approximations

**Butterworth Chebyschev** Elliptic Bessel

Obtained by the LP to BP transformation of the corresponding LP approximations

$$
s \rightarrow \frac{s^2 + 1}{s \cdot BW_{N}}
$$

- Standard LP to BP transform is a variable mapping transform
- Maps jω axis to jω axis
- Maps LP poles to BP poles
- Maps LP zeros to BP zeros
- Preserves basic shape but warps frequency axis
- Doubles order
- Introduces additional zeros at origin (number equal half the order)
- Pole Q of resultant band-pass functions can be very large for narrow pass-band
- Sequencing of frequency scaling and transformation does not affect final function

Example 1: Obtain an approximation that meets the following specifications



 $BW = \omega_{\rm B} - \omega_{\rm A}$ 

$$
\omega_{\rm M} = \sqrt{\omega_{\rm B} \bullet \omega_{\rm A}}
$$

Assume that  $\omega_{AL}$ ,  $\omega_{BH}$  and  $\omega_M$  satsify

$$
\frac{\omega_\text{M}^2\text{-}\omega_\text{AL}^2}{\omega_\text{AL}\text{-BW}}=\frac{\omega_\text{BH}^2\text{-}\omega_\text{M}^2}{\omega_\text{BH}\text{-BW}}
$$

#### **Standard LP to BP Transformation** Recall from above:

Frequency and s-domain Mappings - Denormalized

(subscript variable in LP approximation for notational convenience)





Example 1: Obtain an approximation that meets the following specifications

**(actually ω<sup>A</sup> and ωAL that map to -1 and -ω<sup>S</sup> respectively but show 1 and ω<sup>S</sup> for convenience)** 





 $\mathsf{BW}\mathsf{=}\omega_{_{\mathsf{B}}}\mathsf{-}\omega_{_{\mathsf{A}}}$  $\omega_{_\mathrm{M}}$ = $\sqrt{\omega_{_\mathrm{B}}}$   $\bullet$   $\omega_{_\mathrm{A}}$  In this example,

$$
\frac{\omega_{\scriptscriptstyle\mathrm{M}}^2\text{-}\omega_{\scriptscriptstyle\mathrm{AL}}^2}{\omega_{\scriptscriptstyle\mathrm{AL}}\text{-}\mathrm{BW}}\neq\frac{\omega_{\scriptscriptstyle\mathrm{BH}}^2\text{-}\omega_{\scriptscriptstyle\mathrm{M}}^2}{\omega_{\scriptscriptstyle\mathrm{BH}}\text{-}\mathrm{BW}}
$$







 $\omega_{_{\mathsf{SN}}}^{\vphantom{\dag}}=\mathsf{min}\,\{\omega_{_{\mathsf{S1}}^{\vphantom{\dag}},\omega_{_{\mathsf{S2}}^{\vphantom{\dag}}}\}$  $\mathsf{min}\{\omega_{\scriptscriptstyle \mathsf{S}1},\omega_{\scriptscriptstyle \mathsf{S}2}\}$ 



 $\omega_{_\mathrm{M}}$ = $\sqrt{\omega_{_\mathrm{B}}}$   $\bullet$   $\omega_{_\mathrm{A}}$ 

Example 2: Obtain an approximation that meets the following specifications

 $\omega_{_{\mathsf{SN}}}^{\vphantom{\dag}}=\mathsf{min}\,\{\omega_{_{\mathsf{S1}}^{\vphantom{\dag}},\omega_{_{\mathsf{S2}}^{\vphantom{\dag}}}\}$  $\mathsf{min}\{\omega_{_{\mathsf{S1}}},\!\omega_{_{\mathsf{S2}}}\}$  .

# **Filter Transformations**

Lowpass to Bandpass **(LP to BP)** Lowpass to Highpass **(LP to HP) EXAMPLE 25 LOWPASS to Band-reject** (LP to BR)

- Approach will be to take advantage of the results obtained for the standard LP approximations
- Will focus on flat passband and zero-gain stop-band transformations

#### Flat Passband/Stopband Filters



# **LP to BS Transformation**

Strategy: As was done for the LP to BP approximations, will use a variable mapping strategy that maps the imaginary axis in the s-plane to the imaginary axis in the s-plane so the basic shape is preserved.

$$
X_{IN} \xrightarrow{\text{I}} T_{LPN}(s) \xrightarrow{X_{OUT}} S \longrightarrow f(s) \xrightarrow{X_{IN}} T_{BS}(s) \xrightarrow{X_{OUT}}
$$

$$
f(s) = \frac{\sum_{i=0}^{m_{\text{T}}} a_{\text{T}i} s^{i}}{\sum_{i=0}^{n_{\text{T}}} b_{\text{T}i} s^{i}}
$$

#### **LP to BS Transformation**





Variable Mapping Strategy to Preserve Shape of LP function:

 $\mathsf{F}_\mathsf{N}(\mathsf{s})$  should

map s=0 to s= $\pm$  j $\infty$ map  $s=0$  to  $s=$   $j0$ map s=j1 to s=j $\omega_A$ map s=j1 to s=-j $\omega_B$ map s=-j1 to s=j $\omega_{\rm B}$ map s=-i1 to s=- $i\omega_{\rm A}$ 



map  $\omega$ =0 to  $\omega$  =  $\pm \infty$ map  $\omega$ =0 to  $\omega$  = 0 map  $\omega$ =1 to  $\omega = \omega_A$ map  $\omega$ =1 to  $\omega$ = - $\omega_B$ map  $\omega = -1$  to  $\omega = \omega_B$ map  $\omega = -1$  to  $\omega = -\omega_A$ 





Mapping Strategy: consider variable mapping transform

```
\mathsf{F}_\mathsf{N}(\mathsf{s}) should
map s=0 to s=\pm j\inftymap s=0 to s= j0map s=j1 to s=j\omega_Amap s=j1 to s=-j\omega_{\rm B}map s=-i1 to s=j\omega_{\rm B}map s=-j1 to s=-j\omega_A
```

```
map \omega=0 to \omega = \pm \inftymap \omega=0 to \omega = 0
map \omega=1 to \omega = \omega_Amap \omega=1 to \omega= -\omega_{\rm B}map \omega = -1 to \omega = \omega_Bmap \omega = -1 to \omega = -\omega_A
```
Consider variable mapping

$$
T_{LPN}\left(F_{N}(s)\right) = T_{BSN}\left(s\right)\Big|_{s=\frac{\text{seB}W_{N}}{s^{2}+1}}
$$
\n
$$
s \to \frac{\text{seB}W_{N}}{s^{2}+1}
$$

# Comparison of Transforms



ω

ω

Frequency and s-domain Mappings

(subscript variable in LP approximation for notational convenience)



Un-normalized Frequency and s-domain Mappings

(subscript variable in LP approximation for notational convenience)



Pole Mappings



Can show that the upper hp pole maps to one upper hp pole and one lower hp pole as shown. Corresponding mapping of the lower hp pole is also shown

- Poles lie on a constant-Q line
- Zeros at  $\pm$  j1 (normalized) or at  $\pm j\omega_M$  (un-normalized) of multiplicity n

#### **LP to BS Transformation**



#### Note for γ small, Q<sub>BS</sub> can get very large



$$
s_x \rightarrow \frac{s \bullet BW}{s^2 + \omega_M^2}
$$

- • **Standard LP to BS transformation is a variable mapping transform**
- • **Maps jω axis to jω axis in the s-plane**
- • **Preserves basic shape of an approximation but warps frequency axis**
- • **Order of BS approximation is double that of the LP Approximation**
- • **Pole Q and ω0 expressions are identical to those of the LP to BP transformation**
- • **Pole Q of BS approximation can get very large for narrow BW**
- • **Other variable transforms exist but the standard is by far the most popular**

# **Filter Transformations**

Lowpass to Bandpass **(LP to BP) Lowpass to Highpass** *(LP to HP)* Lowpass to Band-reject **(LP to BR)**

• Approach will also be to take advantage of the results obtained for the standard LP approximations

• Will focus on flat passband and zero-gain stop-band transformations

#### Flat Passband/Stopband Filters



# **LP to HP Transformation**

Strategy: As was done for the LP to BP approximations, will use a variable mapping strategy that maps the imaginary axis in the s-plane to the imaginary axis in the s-plane so the basic shape is preserved.

$$
X_{IN} \longrightarrow T_{LPN}(s) \longrightarrow X_{OUT}
$$
\n
$$
T_{HP}(s) = T_{LPN}(f(s))
$$
\n
$$
T_{HP}(s) = T_{LPN}(f(s))
$$
\n
$$
f(s) = \frac{\sum_{i=0}^{m_T} a_{Ti} s^i}{\sum_{i=0}^{n_T} b_{Ti} s^i}
$$

=0

*i*

#### **LP to HP Transformation**





Mapping Strategy:



Variable Mapping Strategy to Preserve Shape of LP function:



map s=0 to s= $\pm$  j $\infty$ map s=j1 to s=-j1 map  $s = -j1$  to  $s = j1$ 



map  $\omega$ =0 to  $\omega$ = $\infty$ map  $\omega$ =1 to  $\omega$ =-1 map  $\omega$ = -1 to  $\omega$ =1



Mapping Strategy: consider variable mapping transform

 $\mathsf{F}_\mathsf{N}(\mathsf{s})$  should

map s=0 to s= $\pm j$ <sup>∞</sup> map  $s=11$  to  $s=-11$ map  $s = -j1$  to  $s = j1$ 



map  $\omega$ =0 to  $\omega$ = $\infty$ map  $\omega$ =1 to  $\omega$ =-1 map  $\omega = -1$  to  $\omega = 1$ 

Consider variable mapping

$$
T_{\text{LPN}}\left(F(s)\right) = T_{\text{LPN}}\left(s\right)\Big|_{s=\frac{1}{s}}
$$
\n
$$
s \to \frac{1}{s}
$$

# Comparison of Transforms



### **LP to HP Transformation**





Frequency and s-domain Mappings

(subscript variable in LP approximation for notational convenience)



Pole Mappings

Claim: With a variable mapping transform, the variable mapping naturally defines the mapping of the poles of the transformed function



Pole Mappings



Pole Mappings





Highpass poles are scaled in magnitude but make identical angles with imaginary axis

HP pole Q is same as LP pole Q

Order is preserved

(Un-normalized variable mapping transform)





# Stay Safe and Stay Healthy !

# End of Lecture 16